## **IN THE CLAIMS**

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The status of the claims is noted below.

- 1. (Currently Amended) Stereo demultiplexer receiving a frequency demodulated stereo-multiplex signal (m(t)) which comprises at least a stereo-difference signal  $(m_d(t))$ , a stereo-sum signal  $(m_s(t))$  and a pilot carrier, comprising a PLL-circuit (4) to recover the pilot carrier and/or at least one harmonic thereof to perform an amplitude demodulation, characterized in that wherein said PLL-circuit (4) receives the sampling rate decimated stereo-sum signal  $(m_s(t))$  as input signal, which is sampling rate decimated by a decimation factor of D.
- 2. (Currently Amended) Stereo demultiplexer according to claim 1, characterized in that wherein said sampling rate decimated stereo-sum signal (m<sub>s</sub>(t)) is further sampling rate decimated by a decimation factor of E before said PLL-circuit (4) receives it as input signal.
- 3. (Currently Amended) Stereo demultiplexer receiving a frequency demodulated stereo-multiplex signal (m(t)) which comprises at least a stereo-difference signal (m<sub>d</sub>(t)), a stereo-sum signal (m<sub>s</sub>(t)) and a pilot carrier, comprising a PLL-circuit to recover the pilot carrier and/or at least one harmonic thereof to perform an amplitude demodulation, wherein said PLL-circuit receives the sampling rate decimated stereo-sum signal (m<sub>s</sub>(t)) as input signal, which is sampling rate decimated by a decimation factor of D, Stereo-demultiplexer according to claim 1, characterized in that

wherein said PLL-circuit (4) outputs a recovered pilot carrier which is interpolated so that it has a sampling rate equal to that of the frequency demodulated stereo-multiplex signal.

- 4 (Currently Amended) Stereo demultiplexer according to claim 3, wherein characterized in that D-1 or (E·D)-1 interpolated pilot carrier values (y(k/D+1), ..., y(k/D+(D-1))) and one calculated pilot carrier value (y(k/D)) are alternately output.
- 5. (Currently Amended) Stereo demultiplexer according to claim 4, wherein characterized in that said interpolation within the PLL-circuit (4) is performed on basis of a prediction starting at said calculated pilot carrier value.
- 6. (Currently Amended) Stereo demultiplexer according to claim 5, characterized by including
  - a PLL (7, 8, 9, 10, 11, 12) within the PLL-circuit (4) which outputs a phase signal, and
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- a first sinus sine calculation unit (14) which outputs said one calculated pilot carrier value (y(k/D)) on the basis of said phase signal.
- 7. (Currently Amended) Stereo demultiplexer according to claim 6, <del>characterized</del> by including
- second to  $D^{th}$  or  $(E \cdot D)^{th}$  sinus sine calculation units  $(15_1, ..., 15_{D-1})$  each of which outputs one of said D-1 or  $(E \cdot D)$ -1 interpolated pilot carrier values (y(k/D+1), ..., y(k/D+(D-1))) on basis of said phase signal and a respective added phase shift value.
- 8. (Currently Amended) Stereo demultiplexer according to claim 6, <del>characterized</del> by including
- a third multiplexer (13) which multiplies said phase signal with a factor of 2 before it is input to said first sine sinus calculation unit (14) and/or a respective second to  $D^{th}$  or  $(E \cdot D)^{th}$  sinus sine calculation unit  $(15_1, ..., 15_{D-1})$  via a respective second to  $D^{th}$  or  $(E \cdot D)^{th}$  adder  $(16_1, ..., 16_{D-1})$  which adds said respective phase shift value so that the  $2^{nd}$  harmonic of the pilot carrier is generated.
- 9. (Currently Amended) Stereo demultiplexer according to claim 6, wherein eharacterized in that said PLL (7, 8, 9, 10, 11, 12) comprises
- a first multiplier (7) receiving samples of the stereo-sum signal (x(k)) as multiplicant at a first input,
  - a filter (8) receiving the output signal of said first multiplier (7),
- a second multiplier (9) multiplying said output signal of said filter (8) with a PLL gain (PLL loop gain),
- a first adder (11) receiving said output signal of said second multiplier (9) at a first input as a first summand, a constant representing the product of the pilot carrier frequency ( $\omega_{pil}$ ) and the sampling periode(T) at a second input as a second summand, and a delayed phase signal which is the output signal of said first adder (11) at a third input as a third summand,
- a delay element (12) receiving said phase signal of said first adder (11) and supplying said delayed phase signal to said third input of said first adder (11), and
- a <u>cosinus</u> cosine calculation unit (10) receiving the phase signal of said first adder (11) and supplying its output signal as multiplier to a second input of said first multiplier (7).



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